APPLYING SYMMETRICAL COMPONENT TRANSFORM FOR INDUSTRIAL APPLIANCE CLASSIFICATION IN NON-INTRUSIVE LOAD MONITORING

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ABSTRACT

Non-intrusive Load Monitoring (NILM) offers elegant, costeffective, scalable, and load-specific electricity consumption monitoring compared to the traditional way of equipping loads with sensors. NILM techniques have been studied extensively for residential loads. Industrial loads offer challenges for NILM, such as phase imbalance associated with 3-phase lines. Therefore, this work presents a load recognition technique for NILM applying low complexity Fortesque Transform (FT). The FT decomposes the unbalanced 3-phase current waveform extracted from 3-phase aggregate power measurements to balance the given load. The 3-phases current waveform is transformed into an image-like representation using a compressed-euclidean distance matrix to improve the recognition ability further. The image representation is used as input to Convolutional Neural Network (CNN) classifier to learn the patterns of labeled data. Experimental evaluation of the industrial aggregated dataset shows that FT improves recognition performance by 5.8%, compared to the case without FT.

Index Terms— NILM, Industrial Appliances, Three-Phase, Fortesque Transform, Symmetrical Components

1. INTRODUCTION

The rise in industrial energy use is of great concern to climate and sustainability challenges and currently accounts for 40% of global Greenhouse Gas (GHG) emissions [XX]. Hence, understanding the consumption profiles of individual industrial machinery can play an essential role in designing customized energy efficiency and demand management strategies in the industrial sector [1, 2, 3]. For example, detailed information about consumption patterns can help enterprises determine which machines are the most energy-consuming, which ones are faulty, and when to replace or service an old appliance.

Non-intrusive methods (NILM), also known as load disaggregation, is a computation technique that extracts appliance-specific energy consumption profiles from aggregate consumption data monitored at the mains [4]. A critical

step in NILM pipelines is to identify and extract features that are distinctive enough to discriminate across different appliances. In this regard, several approaches have been proposed for residential appliances, with high-frequency current-voltage(V-I) waveforms-based features consistently achieving state-of-the-art performances [5, 6, 7, 8, 9, 10, 11, 12].

Nevertheless, identifying unique and distinct features for appliance recognition can be challenging in an industrial setting with a three-phase power system owing to phase imbalance. In fact, industrial environments often consist of three-phase machines equipped with various equipment such as larger motors and drive [13]. It is also likely to find single-phase appliances that will lead to phase imbalance, as well as the combination of single- and three-phase machines with extensive energy consumption ranges that will make extracting distinctive features for appliance recognition very challenging. The phase imbalance is likely to affect the feature extraction stage of the event-based NILM, especially when the small-consuming appliance has proceeded with larger motors, as illustrated in Figure 1.

There is limited research about using NILM to disaggregate industrial machinery [13], and phase imbalance was not taken into full consideration in the few existing works. The work presented in [14] demonstrates a neural-net-based NILM algorithm for energy disaggregation of a chiller plant system in a commercial building for energy monitoring purposes. In [2] the authors present the FHMM NILM algorithm for energy disaggregation of machinery in a brick production factory. An optimization algorithm for non-intrusive load monitoring of industrial loads in three different industrial sectors: food processing, stonecutting, and glassmaking is presented in [15]. Finally, the work presented in [16], demonstrates that transforming the current waveform into an image-like representation using compressed distance and feeding it as input to a Convolutional Neural Network (CNN) offers higher appliance recognition performance in a threephase system that included industrial loads. Yet, the presented approaches did not exploit the phase imbalance common in an industrial setting.

Against this background, this work exploits the multidimension nature of the unbalanced three-phase system in an industrial setting through low-complexity Fortesque Transform (FT). FT is an instantaneous symmetrical component

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Fig. 1. Current waveform before and after FT transformation when single phase appliance is activated. (a) Current waveform for different events, (b) Current-Voltage trajectory for bulb activation before FT, (c) Current-Voltage trajectory for bulb activation after FT.

[17], a well-known tool for calculating power quality in electrical grids [18]. It has been applied to various problems such as fault location and recognition [19, 17] and in generating the reference currents for the converters in micro-grid systems [20].

In the context of this paper, the FT is used to decompose the unbalanced three-phase current into a set of three components to balance the given load. The base assumption is that decomposing the three-phase current will help in extracting distinctive current features, especially when a small-single phase appliance is preceded by a large three-phase load, as illustrated in Figure 1. For appliance recognition, we follow [16] and transform the balanced 3-phases current waveform into an image-like representation using a compressedeuclidean distance matrix, which is input to the CNN classifier.

2. PROPOSED METHODS

We first present the three-phase system model for the industrial environment used in this work. Then we explain how activation current was obtained from aggregate measurements for single-phase and three-phase data.

A three-phase electrical system widely used in industrial environment consist set of three voltages and currents. In time-domain, the unbalanced three-line voltages and currents can be expressed as:

$$s_1(t) = a_1(t)\cos(2\pi f_0 t + \phi) + \eta_1(t) \tag{1}$$

$$s_2(t) = a_2(t)\cos(2\pi f_0 t + \phi - \frac{2\pi}{3}) + \eta_2(t)$$
 (2)

$$s_3(t) = a_3(t)\cos(2\pi f_0 t + \phi + \frac{2\pi}{3}) + \eta_3(t)$$
(3)

where:

- $s_k(t)$ represent the AC signal either voltage or current,
- a_k(t) is the amplitude of the signal around the fundamental frequency f₀,

- ϕ is the original phase shift of $s_1(t)$, and
- $\eta_k(t)$ represent additive noise present in the signal $s_k(t)$

In an ideal situation, the three-phase system is assumed to be balanced such that the three quantities have the same amplitude $(a_1(t) = a_2(t) = a_3(t))$ and an equal phase shift of $\frac{2\pi}{3}$. However, in industrial setting the balance assumption usually does not hold. As the result, a symmetrical component transform such as FT will decompose the unbalanced threephase system into balanced and unbalanced components, namely positive, negative and zero-sequence components [21].

Given a three-phase signal $\mathbf{S}(t) = [s_1(t), s_2(t), s_3(t)]^T$, the transformed FT components $\mathbf{S}_{+,-,0}$ are given by

$$\mathbf{S}_{+,-,0} = \begin{bmatrix} s_+(t)\\ s_-(t)\\ s_0(t) \end{bmatrix} = \frac{2}{3} \mathbf{F} \mathbf{s}(t) \tag{4}$$

where:

$$\mathbf{F} = \begin{bmatrix} 1 & a & a^2 \\ 1 & a^2 & a \\ 1 & 1 & 1 \end{bmatrix}$$
(5)

where $a = e^{j\frac{2}{3}}$.

For balanced three-phase systems only the $s_+(t)$ and $s_-(t)$ sequence exist, while the zero-sequence $s_0(t)$ will be null. Thus the zero-sequence $s_0(t)$ can quantify the amount of imbalance in the three-phase signal. We utilize this hypothesis and assume that the zero component $s_0(t)$ will represent the appliance signature of a single-phase appliance in case of phase imbalance as shown in Figure 1c.

2.1. Feature Extraction and Pre-Processing from Aggregate Measurements

We consider appliance features extracted in brief time windows to recognize appliances, containing only one event feature derived from three-phase aggregate power measurements. Thus an activation current i and voltage v is obtained by first measuring 25 complete cycles of current and voltage before $\{\mathbf{i}_k^{(b)}, \mathbf{v}_k^{(b)}\}$ and after $\{\mathbf{i}_k^{(a)}, \mathbf{v}_k^{(a)}\}$ the event, where $k \in \{1, 2, 3\}$. The 25 cycles correspond to steady-state behaviour and are equivalent to $T_s \times N_s$ samples where $T_s = \frac{f_s}{f}$, $f_s = 50 KHz$ is sampling frequency and f = 100 Hz is the mains frequency.

The 25 cycles are aligned at zero-crossing of the voltage and thereafter, one-cycle activation current before $i_k^{(b)}$ and after $i_k^{(a)}$ event is extracted. The activation current *i* is then calculated as follows: $i_k = i_k^{(a)} - i_k^{(b)}$ and $v_k = v_k^{(a)}$ if the event is caused by activation of appliance, and $i_k = i_k^b - i_k^{(a)}$ if the event is caused by de-activation of appliance.

2.2. Compressed Euclidean Distance Matrix

The Euclidean Distance Matrix (EDM) is the matrix of squared Euclidean distances representing the spacing of a set of w points in euclidean space [22] such that:

$$D_{w,w} = \begin{bmatrix} 0 & d_{1,2} & \cdots & d_{1,w} \\ d_{2,1} & 0 & \cdots & d_{2,w} \\ \vdots & \vdots & \ddots & \vdots \\ d_{w,1} & d_{w,2} & \cdots & 0 \end{bmatrix}$$
(6)

where $d_{u,v} = ||i(t)_u - i(t)_v||_2$ is the Euclidean distance function which is is widely used as a pre-processing step for many machine learning approaches such as K-means clustering and K-nearest neighbor algorithms [23, 22]. The compressed EDM provides a relationship metric between each element in the time series. It encodes time series data into a reccurence graph (RP) which reveal in which point trajectories return to a previous state. It is usually formulated as;

$$RG_{i,j} = \begin{cases} \delta & \text{if } \tau > \delta \\ \tau & \text{otherwise} \end{cases}$$
(7)

where $\tau = \left\lfloor \frac{d_{u,v}}{\epsilon} \right\rfloor, 0 < \epsilon \le 1$ and $\delta \ge 1$

Thus to generate RG images of size $w \times w$, the piecewise aggregation approximation (PAA) [24, 1], is first applied to reduce the signal's dimension from T_s to a predefined size wwith minimal information loss. The embedding size w is the hyper-parameter that needs to be selected in advance. Previous works confirmed that the choice of w does not significantly influence the classification performance. Yet, large values of w impact the learning speed, and very small values of w will most likely lead to larger information loss [1, 16]. The PAA signal is then transformed into an RP image using Equation (7).

2.3. Classifier and Training Procedure

Once the appliance features have been extracted, a generic machine-learning classifier can learn the labeled data pattern.

A 2-stage CNN is used for this task. Each CNN layer contains 50 feature maps with a filter size of 5×5 , stride size of 1×1 , and ReLU activation function followed by a pooling layer with a filter size of 2×2 . The final layer consists of two FC layers with a hidden dimension of w^2 and K, respectively, where K is the number of classes determined by the number of appliances available. The final predicted class is obtained by applying softmax activation function. Since the problem at hand is multi-class classification the Cross-Entropy Loss function defined in Equation (8) is used.

$$\mathcal{L}_{\theta}(y,p) = -\sum_{i=1}^{M} y_i \cdot \log p_i \tag{8}$$

The CNN network is trained for 600 iterations using minibatch Stochastic Gradient Descent (SDG) with a momentum of 0.9, a learning rate of 1^{-3} , and a batch size of 16. To avoid over-fitting early stopping with patience is used where the training model is terminated once the validation performance does not change after 20 iterations.

2.4. Evaluation Procedure

The proposed method is evaluated on Laboratory-measured Industrial Load of Appliance Characteristics dataset (LILACD) [25]. The LILACD dataset contains three-phase aggregated and sub-metered current and voltage measurements sampled at 50 kHz for 16 different appliance types (industrial and home appliances). In this study, aggregated measurement data that contain measurements of more than one concurrently running appliance is used. Since the LILACD dataset is multi-dimensional, a separate activation current for each current phase is an input feature.

We adopt stratified 4-fold cross-validation with a random shuffle to benchmark our approach. The 4-fold crossvalidation with a random shuffle provides stratified randomized folds while preserving the label's percentage in each fold. We analyzed the FT influence by assessing the recognition appliance performance with and without FT transform. This was achieved by training the CNN-based classifier with RG and V-I feature representation. The V-I was generated using the same process for generating binary V-I described in [7, 1].

We quantitatively evaluate the classification performance with macro averaged F_1 (%) score is defined as $F_{macro} = 100 \cdot \frac{1}{M} \sum_{i=1}^{M} F_1^{(i)}$ where M is the number of appliances and F_1 is the harmonic mean of precision and recall. We also use the confusion matrix, which shows the correct predictions (the diagonal) and provides a clear view of which appliances are confused with each other.

3. RESULTS AND DISCUSSION

Figure 2 presents the per-appliance F-score for the two feature representations with and without FT transform. In both cases,

we see that the RP achieves a high F-score compared to V-I. We also see that when FT is used, the RP performs overall 96.43 scores compared to 94.21 when FT is not applied. We further see in Figure 2b that the RP with FT attains an Fscore above 0.9 for all appliances. Notably, the V-I achieve a low score of less than 60 for single-phase appliances such as Coffe-machine, Kettle, Raclette, and Hairdryer.



Fig. 2. Classification Results: a) F-score combined phases without FT for VI and RP b) F-score combined phases with FT for VI and RP.

Investigating the confusion matrix in Figure 3, we observe that the RP without FT makes several confusions between Hair-dryer and Raclette. We also see that applying FT improves the recognition performance of most single-phase appliances. The results suggest that the proposed FT transform can improve the process of extracting appliance signatures that is distinct enough to recognize appliances especially when there is a phase imbalance, as initially hypothesized.



Fig. 3. Confusion Matrix: (a) RP without FT (b) RP with FT.

4. CONCLUSION

In this work, we have applied low-complexity Fortesque Transform to improve the discrimination power of NILM for industrial machinery recognition. The FT decomposes the unbalanced three-phase current waveform into a set of three components to balance the given load. The decomposed current signal was then transformed into an image-like representation using the euclidean-distance-similarity function (RG) and fed into the CNN for classification.

We then present the outcome of an empirical investigation on how FT influences the classification performance for RP and V-I features. We observe that applying FT improves classification performances for both V-I and RP features. The outcomes further suggest that the RP feature with FT is consistently superior to those from V-I images with FT. Most importantly, RP with FT can identify single-phase appliances whose signatures have been distorted by phase imbalance and high-consuming appliances. Ultimately, these results suggest that industries could effectively use the proposed approach to disaggregate industrial machinery and allow value-added services such as predictive maintenance by monitoring power consumption from a single point source.

Finally, while the results argue in favor of the proposed approach, it is important to acknowledge that the experimental procedure is limited to one single dataset. Hence, in order to better generalize the findings, further experimentation is required in other datasets. Furthermore, it is important to remark that the developed method assumes perfect detection of the appliance change of state (i.e., from ON to OFF and vice-versa), which may be hard to achieve especially in an industrial setting where the aggregated signal tends to be far more complex than for example in residential spaces. Therefore, future work iterations should also consider the effect of appliance state change detection in the overall non-intrusive appliance identification process.

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